

October 1987 Revised January 1999

CD4046BC

Micropower Phase-Locked Loop

General Description

The CD4046BC micropower phase-locked loop (PLL) consists of a low power, linear, voltage-controlled oscillator (VCO), a source follower, a zener diode, and two phase comparators. The two phase comparators have a common signal input and a common comparator input. The signal input can be directly coupled for a large voltage signal, or capacitively coupled to the self-biasing amplifier at the signal input for a small voltage signal.

Phase comparator I, an exclusive OR gate, provides a digital error signal (phase comp. I Out) and maintains 90° phase shifts at the VCO center frequency. Between signal input and comparator input (both at 50% duty cycle), it may lock onto the signal input frequencies that are close to harmonics of the VCO center frequency.

Phase comparator II is an edge-controlled digital memory network. It provides a digital error signal (phase comp. II Out) and lock-in signal (phase pulses) to indicate a locked condition and maintains a 0° phase shift between signal input and comparator input.

The linear voltage-controlled oscillator (VCO) produces an output signal (VCO Out) whose frequency is determined by the voltage at the VCO $_{\rm IN}$ input, and the capacitor and resistors connected to pin C1 $_{\rm A}$, C1 $_{\rm B}$, R1 and R2.

The source follower output of the VCO $_{IN}$ (demodulator Out) is used with an external resistor of 10 k Ω or more.

The INHIBIT input, when high, disables the VCO and source follower to minimize standby power consumption. The zener diode is provided for power supply regulation, if necessary.

Features

■ Wide supply voltage range: 3.0V to 18V

 \blacksquare Low dynamic power consumption: 70 μW (typ.) at f_0 = 10 kHz, V_{DD} = 5V

■ VCO frequency: 1.3 MHz (typ.) at $V_{DD} = 10V$

■ Low frequency drift: 0.06%/°C at V_{DD} = 10V with temperature

■ High VCO linearity: 1% (typ.)

Applications

- · FM demodulator and modulator
- · Frequency synthesis and multiplication
- · Frequency discrimination
- · Data synchronization and conditioning
- Voltage-to-frequency conversion
- · Tone decoding
- FSK modulation
- Motor speed control

Ordering Code:

Order Number	Package Number	Package Description
CD4046BCM	M16A	16-Lead Small Outline integrated Circuit (SOIC), JEDEC MS-012, 0.150" Narrow Body
CD4046BCN	N16E	16-Lead Plastic Dual-In-Line Package (PDIP), JEDEC MS-001, 0.300" Wide

Devices also available in Tape and Reel. Specify by appending the suffix letter "X" to the ordering code.

Connection Diagram

PIN Assignments for SOIC and DIP PHASE PULSES 1 PHASE COMP I OUT 2 COMPARATOR IN 3 VCO OUT 4 INHIBIT 5 C1_A 6 C1_B 7 V_{SS} 8 Top View

CD4046BC Block Diagram COMPARATOR PHASE COMP II OUT DEMODULATOR 10 OUT ZENER FIGURE 1.

Absolute Maximum Ratings(Note 1)

(Note 2)

 $\begin{array}{ll} \text{DC Supply Voltage (V}_{\text{DD}}) & -0.5 \text{ to } +18 \text{ V}_{\text{DC}} \\ \text{Input Voltage (V}_{\text{IN}}) & -0.5 \text{ to } \text{V}_{\text{DD}} +0.5 \text{ V}_{\text{DC}} \\ \text{Storage Temperature Range (T}_{\text{S}}) & -65^{\circ}\text{C to } +150^{\circ}\text{C} \end{array}$

Power Dissipation (P_D)

 Dual-In-Line
 700 mW

 Small Outline
 500 mW

Lead Temperature (T_L)

(Soldering, 10 seconds) 260°C

Recommended Operating Conditions (Note 2)

DC Supply Voltage (V_{DD}) 3 to 15 V_{DC} Input Voltage (V_{IN}) 0 to V_{DD} V_{DC} Operating Temperature Range (T_{A}) -40° C to $+85^{\circ}$ C

Note 1: "Absolute Maximum Ratings" are those values beyond which the safety of the device cannot be guaranteed. They are not meant to imply that the devices should be operated at these limits. The table of "Recommended Operating Conditions" and "Electrical Characteristics" provides conditions for actual device operation.

Note 2: $V_{SS} = 0V$ unless otherwise specified.

DC Electrical Characteristics (Note 2)

0	Parameter	O and distance	-40	-40°C		+25°C			+85°C	
Symbol	Parameter	Conditions	Min	Max	Min	Тур	Max	Min	Max	Units
I _{DD}	Quiescent Device Current	Pin 5 = V_{DD} , Pin 14 = V_{DD} ,								
		Pin 3, 9 = V_{SS}								
		$V_{DD} = 5V$		20		0.005	20		150	μΑ
		$V_{DD} = 10V$		40		0.01	40		300	μΑ
		V _{DD} = 15V		80		0.015	80		600	μΑ
		Pin 5 = V _{DD} , Pin 14 = Open,								
		Pin 3, $9 = V_{SS}$								
		$V_{DD} = 5V$		70		5	55		205	μΑ
		$V_{DD} = 10V$		530		20	410		710	μΑ
		$V_{DD} = 15V$		1500		50	1200		1800	μΑ
V _{OL}	LOW Level Output Voltage	$V_{DD} = 5V$		0.05		0	0.05		0.05	V
		V _{DD} = 10V		0.05		0	0.05		0.05	V
		$V_{DD} = 15V$		0.05		0	0.05		0.05	V
V _{OH}	HIGH Level Output Voltage	$V_{DD} = 5V$	4.95		4.95	5		4.95		V
		$V_{DD} = 10V$	9.95		9.95	10		9.95		V
		$V_{DD} = 15V$	14.95		14.95	15		14.95		V
V _{IL}	LOW Level Input Voltage	$V_{DD} = 5V$, $V_{O} = 0.5V$ or 4.5V		1.5		2.25	1.5		1.5	V
	Comparator and Signal In	$V_{DD} = 10V$, $V_{O} = 1V$ or $9V$		3.0		4.5	3.0		3.0	V
		$V_{DD} = 15V$, $V_{O} = 1.5V$ or $13.5V$		4.0		6.25	4.0		4.0	V
V _{IH}	HIGH Level Input Voltage	$V_{DD} = 5V, V_{O} = 0.5V \text{ or } 4.5V$	3.5		3.5	2.75		3.5		V
	Comparator and Signal In	$V_{DD} = 10V, V_{O} = 1V \text{ or } 9V$	7.0		7.0	5.5		7.0		V
		$V_{DD} = 15V$, $V_{O} = 1.5V$ or $13.5V$	11.0		11.0	8.25		11.0		V
I _{OL}	LOW Level Output Current	$V_{DD} = 5V, V_{O} = 0.4V$	0.52		0.44	0.88		0.36		mA
	(Note 4)	$V_{DD} = 10V, V_{O} = 0.5V$	1.3		1.1	2.25		0.9		mA
		$V_{DD} = 15V, V_{O} = 1.5V$	3.6		3.0	8.8		2.4		mA
I _{OH}	HIGH Level Output Current	$V_{DD} = 5V, V_{O} = 4.6V$	-0.52		-0.44	-0.88		-0.36		mA
	(Note 4)	$V_{DD} = 10V, V_{O} = 9.5V$	-1.3		-1.1	-2.25		-0.9		mA
		$V_{DD} = 15V, V_{O} = 13.5V$	-3.6		-3.0	-8.8		-2.4		mA
I _{IN}	Input Current	All Inputs Except Signal Input								
		$V_{DD} = 15V, V_{IN} = 0V$		-0.3		-10^{-5}	-0.3		-1.0	μΑ
		$V_{DD} = 15V, V_{IN} = 15V$		0.3		10 ⁻⁵	0.3		1.0	μΑ
C _{IN}	Input Capacitance	Any Input (Note 3)					7.5			pF
P _T	Total Power Dissipation	$f_0 = 10 \text{ kHz}, R1 = 1 \text{ M}\Omega,$								
		$R2 = \infty, \varsigma X O_{IN} = \varsigma_{\Delta\Delta}/2$								
		$V_{DD} = 5V$				0.07				mW
		$V_{DD} = 10V$				0.6				mW
		$V_{DD} = 15V$				2.4				mW

Note 3: Capacitance is guaranteed by periodic testing.

Note 4: \mathbf{I}_{OH} and \mathbf{I}_{OL} are tested one output at a time.

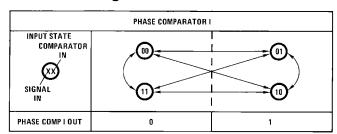
Symbol	Parameter	Conditions	Min	Тур	Max	Units
VCO SECT	ION					1
DD	Operating Current	$f_0 = 10 \text{ kHz}, R1 = 1 \text{ M}\Omega,$				
55		$R2 = \infty$, $\varsigma XO_{IN} = \varsigma_{\Lambda\Lambda}/2$				
		$V_{DD} = 5V$		20		μА
		$V_{DD} = 10V$		90		μA
		V _{DD} = 15V		200		μA
f _{MAX}	Maximum Operating Frequency	C1 = 50 pF, R1 = 10 k Ω ,				
		$R2 = \infty$, $\zeta XO_{IN} = \zeta_{\Delta\Delta}$				
		$V_{DD} = 5V$	0.4	0.8		MHz
		V _{DD} = 10V	0.6	1.2		MHz
		V _{DD} = 15V	1.0	1.6		MHz
	Linearity	VCO _{IN} = 2.5V ±0.3V,				
		$R1 \ge 10 \text{ k}\Omega, V_{DD} = 5V$		1		%
		VCO _{IN} = 5V ±2.5V,				
		$R1 \ge 400 \text{ k}\Omega, V_{DD} = 10V$		1		%
		VCO _{IN} = 7.5V ±5V,				
		$R1 \ge 1 M\Omega$, $V_{DD} = 15V$		1		%
	Temperature-Frequency Stability	%/°C∝1/φ. ς _{ΔΔ}				
	No Frequency Offset, f _{MIN} = 0	R2 = ∞				
		$V_{DD} = 5V$		0.12-0.24		%/°C
		V _{DD} = 10V		0.04-0.08		%/°C
		V _{DD} = 15V		0.015-0.03		%/°C
	Frequency Offset, f _{MIN} ≠ 0	$V_{DD} = 5V$		0.06-0.12		%/°C
		V _{DD} = 10V		0.05-0.1		%/°C
		$V_{DD} = 15V$		0.03-0.06		%/°C
VCO _{IN}	Input Resistance	$V_{DD} = 5V$		10 ⁶		МΩ
		$V_{DD} = 10V$		10 ⁶		$M\Omega$
		$V_{DD} = 15V$		10 ⁶		MΩ
VCO	Output Duty Cycle	$V_{DD} = 5V$		50		%
		$V_{DD} = 10V$		50		%
		$V_{DD} = 15V$		50		%
t _{THL}	VCO Output Transition Time	$V_{DD} = 5V$		90	200	ns
t _{THL}		$V_{DD} = 10V$		50	100	ns
		$V_{DD} = 15V$		45	80	ns
PHASE CO	MPARATORS SECTION					
R _{IN}	Input Resistance					
	Signal Input	$V_{DD} = 5V$	1	3		МΩ
		$V_{DD} = 10V$	0.2	0.7		МΩ
		$V_{DD} = 15V$	0.1	0.3		MΩ
	Comparator Input	$V_{DD} = 5V$		10 ⁶		МΩ
		$V_{DD} = 10V$		10 ⁶		МΩ
		$V_{DD} = 15V$		10 ⁶		МΩ
	AC-Coupled Signal Input Voltage	C _{SERIES} = 1000 pF				
	Sensitivity	f = 50 kHz				
		$V_{DD} = 5V$		200	400	mV
		$V_{DD} = 10V$		400	800	mV
		V _{DD} = 15V		700	1400	mV

AC Electrical Characteristics (Continued)

Symbol	Parameter	Conditions	Min	Тур	Max	Units
DEMODUL	ATOR OUTPUT	•	•		•	•
VCO _{IN} -	Offset Voltage	$RS \ge 10 \ k\Omega, \ V_{DD} = 5V$		1.50	2.2	V
V _{DEM}		$RS \geq 10 \ k\Omega, \ V_{DD} = 10 V$		1.50	2.2	V
		$RS \geq 50~k\Omega,~V_{DD} = 15V$		1.50	2.2	V
	Linearity	RS ≥ 50 kΩ				
		$VCO_{IN} = 2.5V \pm 0.3V$, $V_{DD} = 5V$		0.1		%
		$VCO_{IN} = 5V \pm 2.5V, V_{DD} = 10V$		0.6		%
		$VCO_{IN} = 7.5V \pm 5V, V_{DD} = 15V$		0.8		%
ZENER DIC	DDE					
VZ	Zener Diode Voltage	$I_Z = 50 \mu A$	6.3	7.0	7.7	V
R _Z	Zener Dynamic Resistance	I _Z = 1 mA		100		Ω

Note 5: AC Parameters are guaranteed by DC correlated testing.

Phase Comparator State Diagrams



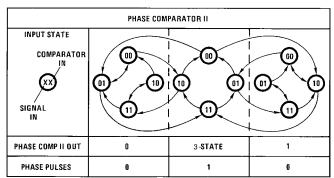
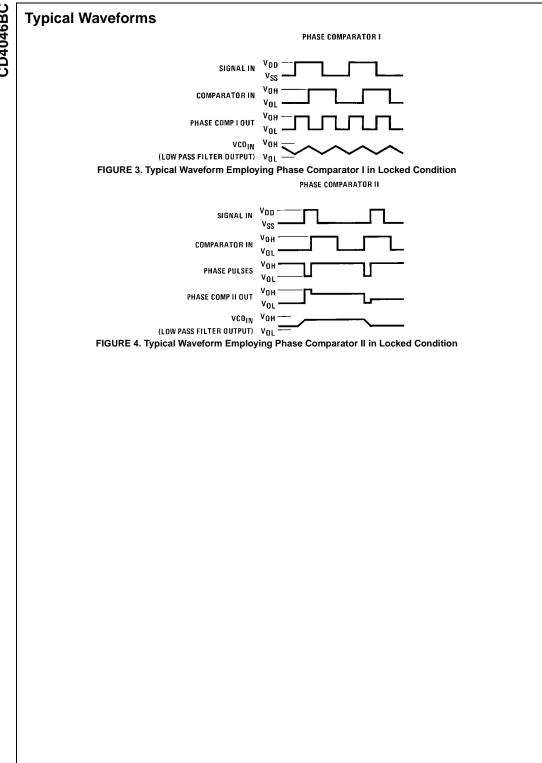
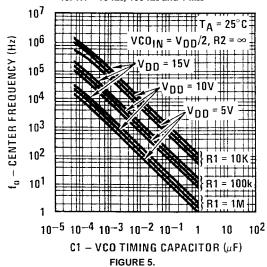


FIGURE 2.

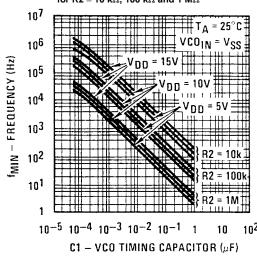


Typical Performance Characteristics

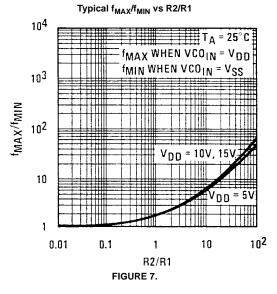
Typical Center Frequency vs C1 for R1 = 10 k Ω , 100 k Ω and 1 M Ω



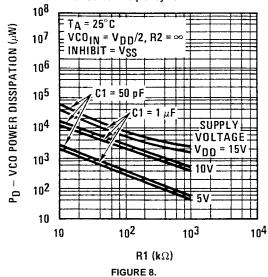
Typical Frequency vs C1 for R2 = 10 k Ω , 100 k Ω and 1 M Ω



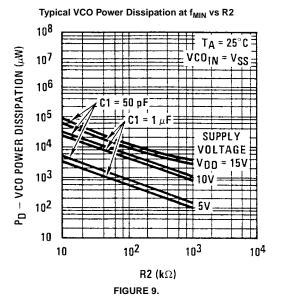
Note: To obtain approximate total power dissipation of PLL system for no-signal input: Phase Comparator I, P_D (Total) = P_D (f_0) + P_D (f_{MIN}) + P_D (R_S); Phase Comparator II, P_D (Total) = P_D (f_{MIN}).



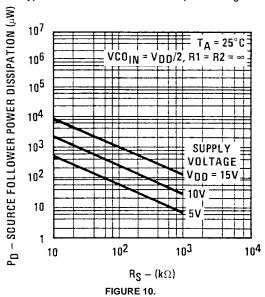
Typical VCO Power Dissipation at Center Frequency vs R1



Note: To obtain approximate total power dissipation of PLL system for no-signal input: Phase Comparator I, P_D (Total) = P_D (f_{O}) + P_D (f_{MIN}) + P_D (R_S); Phase Comparator II, P_D (Total) = P_D (f_{MIN}).



Typical Source Follower Power Dissipation vs R_S



Note: To obtain approximate total power dissipation of PLL system for no-signal input: Phase Comparator I, P_D (Total) = P_D (f_{O}) + P_D (f_{MIN}) + P_D (R_S); Phase Comparator II, P_D (Total) = P_D (f_{MIN}).

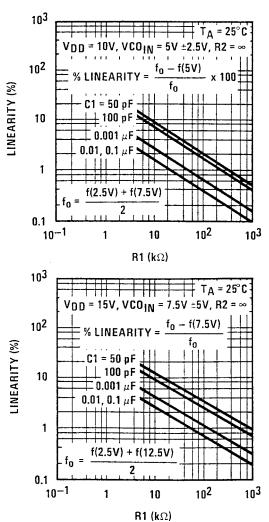


FIGURE 11. Typical VCO Linearity vs R1 and C1

Note: To obtain approximate total power dissipation of PLL system for no-signal input: Phase Comparator I, P_D (Total) = P_D (f_{O}) + P_D (f_{MIN}) + P_D (R_S); Phase Comparator II, P_D (Total) = P_D (f_{MIN}).

Design Information

This information is a guide for approximating the value of external components for the CD4046B in a phase-locked-loop system. The selected external components must be within the following ranges: R1, R2 \geq 10 k Ω , R_S \geq 10 k Ω , C1 \geq 50 pE

In addition to the given design information, refer to Figure 5, Figure 6, Figure 7 for R1, R2 and C1 component selections.

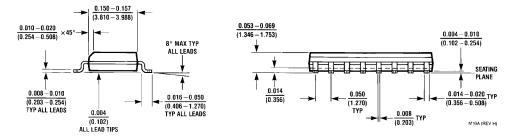
	Using Phase	Comparator I	Using Phase Comparator II			
Characteristics	VCO Without Offset	VCO With Offset	VCO Without Offset	VCO With Offset		
	R2 = ∞		R2 = ∞			
VCO Frequency	IMAX 10 21L VDD/2 VDD VCO INPUT VOLTAGE	MAX fo 211 YOD/2 VDD VCD INPUT VOLTAGE	IMAX To 2 IL VDD/2 VDD VCO INPUT VOLTAGE	1 NAX 2 1 1 2 1 1 1 1 1 1 1 1 1 1 1 1 1 1 1		
For No Signal Input	VCO in PLL sy	stem will adjust	VCO in PLL sys	tem will adjust to		
	to center fr	equency, f _o	lowest operating	g frequency, f _{min}		
Frequency Lock		2 f _L = full VCO f	frequency range			
Range, 2 f _L		$2 f_L = f_m$	$_{\text{max}} - f_{\text{min}}$			
Frequency Capture Range, 2 f _C	11 × R3 C2 = C2	$2 f_{\text{C}} pprox rac{1}{\pi} \sqrt{rac{2 \pi f_{\text{L}}}{ au_{1}}}$				
Loop Filter Component Selection	1N 0 0 0 U T	For 2 f _C , see Ref.	f _C :	= f _L		
Phase Angle Between	90° at center frequen	cy (f _o), approximating	Always (0° in lock		
Single and Comparator	0° and 180° at ends	s of lock range (2 f _L)				
Locks on Harmonics	Ye	es	N	lo		
of Center Frequency						
Signal Input Noise	Hi	gh	Lo	OW		
Rejection						

	Using Phase	Comparator I	Using Phase Comparator II			
Characteristics	VCO Without Offset	VCO With Offset	VCO Without Offset	VCO With Offset		
	R2 = ∞		R2 = ∞			
VCO Component	Given: f _o .	Given: fo and fL.	Given: f _{max} .	Given: f _{min} and f _{max} .		
Selection	Use fo with	Calculate f _{min}	Calculate fo from	Use f _{min} with		
	Figure 5 to	from the equation	the equation	Figure 6 to		
	determine R1 and C1.	$f_{min} = f_o - f_L$.	$f_0 = \frac{f_{max}}{2}$.	to determine R2 and C1.		
		Use f _{min} with Figure 6 to determine R2 and C1.		Calculate $\frac{f_{max}}{f_{min}}$		
			Use fo with Figure 5 to	17/111		
		Calculate	determine R1 and C1.	Use		
		f _{max} f _{min}		f _{max} f _{min} with Figure 7		
		from the equation		to determine ratio		
		$\frac{f_{max}}{f_{min}} = \frac{f_0 + f_L}{f_0 - f_L}.$ Use		R2/R1 to obtain R1.		
		f _{max} f _{min} with Figure 7				
		to determine ratio R2/				
		R1 to obtain R1.				

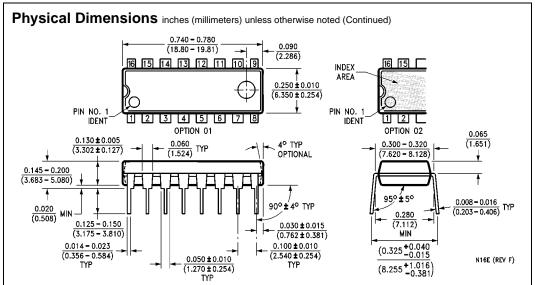
References

G.S. Moschytz, "Miniaturized RC Filters Using Phase-Locked Loop", BSTJ, May, 1965. Floyd Gardner, "Phaselock Techniques", John Wiley & Sons, 1966.

LEAD NO.1



16-Lead Small Outline Integrated Circuit (SOIC), JEDEC MS-012, 0.150" Narrow Body Package Number M16A



16-Lead Plastic Dual-In-Line Package (PDIP), JEDEC MS-001, 0.300" Wide Package Number N16E

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